Rough surface scattering at low-grazing incidence: A dedicated model

M. Saillard¹ and G. Soriano²

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[1] The rough surface radar cross section becomes very small at grazing incidence, since most part of the energy travels in the forward direction. We build a rigorous model that can estimate it accurately. It is based on a specific integral formalism, and applies to both low-grazing incidence and scattering angles. Reference data are produced to test the validity of approximate models, and a technique for extrapolation is detailed.

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1. Introduction

[2] The way to compute the non-grazing radar cross section (RCS) from a random rough surface separating two homogeneous media without any simplifying assumption appears in the work of *Wagner et al.* [1997] and is detailed by *Tsang et al.* [2001]. It is based on the numerical solution of a set of boundary integral equations [*Poggio and Miller*, 1973; *Colton and Kress*, 1983] for a deterministic rough surface of finite area. A plane wave cannot be the incident field, because of edge effects: a tapered polarized beam [*Toporkov et al.*, 1999; *Braunisch et al.*, 2000] with a Gaussian-like footprint on the surface is the reference method in order to bound to the finite area of the rough surface the numerical support of the integral equations unknowns, namely the tangential components of the total electromagnetic field. Statistical RCS is obtained through Monte Carlo average.

[3] The RCS from a rough surface becomes very small at low-grazing incidence (LGI), since most part of the energy travels in the forward direction. In the work of Spiga et al. [2008], we have presented the Grazing Method of Moments (GMoM), a different but rigorous approach dedicated to the time harmonic electromagnetic wave scattering from rough surfaces under grazing illumination. The rough surface is represented by a bounded perturbation of a plane and is enlightened by a plane wave. The integral equations are built on specific unknowns and right hand sides that permit edge effects to be avoided. In the work of Soriano et al. [2010], the implementation of the GMoM for two-dimensional problems is detailed, and numerical results on one-dimensional (1D) Gaussian surfaces in Optics and 1D ocean-like surfaces at microwave frequencies are shown. Note that several other methods have been considered to eliminate the tapered wave requirement in rough surface scattering. They are summarized in the introduction of Spiga et al. [2008]. Plane wave

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illuminating a bounded perturbation of the plane appears in the work of *West* [2003] and *Zhao et al.* [2005], but in those papers, edge effects are avoided thanks to resistive loadings. Also, in the work of *Zhao et al.* [2005], an image theory involving the half plane Green's function is addressed. The GMoM differs from those two pioneering approaches in that sense that first, the physical boundary condition on the rough surface is left untouched, and second, it only makes use of classical integral operators, built on the free space Green's function.

[4] This paper is organized as follows. Below this introduction, the boundary integral formalism is presented in section 2: after classical theory is recalled, integral equations are modified for numerical solution with an incident plane wave. Also, lossy dielectrics and conductors dielectric are given a specific treatment through an impedance boundary condition. In spite of appearances, the scattered field theoretically shows at both LGI and low-grazing scattering angles (LGS) a very simple and general behavior. This behavior, and how to enforce it numerically, is underlined in section 3, while in section 4, a technique for extrapolation, from one low-grazing angle to another, is suggested. Then in section 5, the GMoM is applied to two-dimensional (2D) perfectly conducting surfaces in the HH polarization case (both the incident electric field and the scattered electric field of interest are normal to the incidence plane). New numerical results are presented for 2D surfaces with Gaussian and ocean spectra.

2. Boundary Integral Formalism

[5] The study is time harmonic at pulsation ω , with a time dependence $e^{-i\omega t}$ assumed. We start by considering an infinite rough surface separating vacuum as upper medium from a semi-infinite dielectric homogeneous lower medium, illuminated by a wave impinging from vacuum. The surface Σ is a perturbation of the (xOy) plane in the right Cartesian coordinate (x, y, z) system with the *z* axis directed toward vacuum (Figure 1). The surface height is given by equation $z = h(\mathbf{r}) = h(x, y)$, and $\hat{\mathbf{n}}$ denotes its unit normal vector directed upward. If θ_0 denotes the incidence angle, counted as indicated on Figure 1, then the incident field is low grazing when

¹Laboratoire de Sondage Electromagnétique de l'Environnement Terrestre, Université du Sud Toulon-Var, La Garde, France.

²Sondage ElectroMagnétique et Optique, Institut Fresnel, Aix-Marseille Université, Marseille, France.



Figure 1. The bounded perturbation of the plane model for low-grazing incidence angles wave scattering from rough surfaces.

 $\theta_0 \rightarrow \pm \pi/2$. The grazing angle can be defined as $\gamma_0 = \frac{\pi}{2} - |\theta_0| \rightarrow 0$ but we prefer to consider $q_0 = K_0 \cos \theta_0 = K_0 \sin \gamma_0$ that shares the same behavior. $K_0 = \omega \sqrt{\varepsilon_0 \mu_0} = \frac{2\pi}{\lambda}$ is the wave number for the vacuum and q_0 is the positive *z* component of the incident wave vector $\mathbf{k}_0 - q_0 \hat{\mathbf{z}}$. At the end, we are mainly concerned by the far field. Denoting $\mathbf{k} + q\hat{\mathbf{z}}$ the scattering wave vector in air at angle θ counted from normal or $\gamma = \pi/2 - |\theta|$ from grazing, the LGS limit corresponds to $\gamma \rightarrow 0$ or $q = K_0 \cos \theta = K_0 \sin \gamma \rightarrow 0$.

[6] Following the classical Stratton-Chu theory [*Poggio* and Miller, 1973; Colton and Kress, 1983], the tangential components of the total fields on the rough surface $\hat{\mathbf{n}} \times \mathbf{E}$ and $\hat{\mathbf{n}} \times \mathbf{H}$ are the solution of integral equations for the air

$$\left(\frac{1}{2} + M_0\right)\hat{\mathbf{n}} \times \mathbf{H} - \frac{i}{\omega\mu_0} P_0 \,\hat{\mathbf{n}} \times \mathbf{E} = \hat{\mathbf{n}} \times \mathbf{E}^i \tag{1}$$

$$\left(\frac{1}{2} + M_0\right)\hat{\mathbf{n}} \times \mathbf{E} + \frac{i}{\omega\varepsilon_0} P_0 \,\hat{\mathbf{n}} \times \mathbf{H} = \hat{\mathbf{n}} \times \mathbf{H}^i \tag{2}$$

and for the scattering medium

$$\left(\frac{1}{2} - M\right)\hat{\mathbf{n}} \times \mathbf{H} + \frac{i}{\omega\mu_0} P \,\hat{\mathbf{n}} \times \mathbf{E} = \mathbf{0} \tag{3}$$

$$\left(\frac{1}{2} - M\right)\hat{\mathbf{n}} \times \mathbf{E} - \frac{i}{\omega\varepsilon_0\varepsilon_r}P\,\hat{\mathbf{n}} \times \mathbf{H} = \mathbf{0} \tag{4}$$

 M_0 and P_0 are the integral operators as defined by *Martin and Ola* [1993]. They write for a tangential density **c**

$$M_0 \mathbf{c}(\mathbf{R}) = \hat{\mathbf{n}} \times \mathbf{curl} \int_{\Sigma} G_0 \mathbf{c}(\mathbf{R}') dS_{\mathbf{R}'}$$
(5)

$$P_0 \mathbf{c}(\mathbf{R}) = \hat{\mathbf{n}} \times \mathbf{curl} \, \mathbf{curl} \int_{\Sigma} G_0 \, \mathbf{c}(\mathbf{R}') dS_{\mathbf{R}'} \tag{6}$$

with $\mathbf{R}' = \mathbf{r}' + h(\mathbf{r}')\hat{\mathbf{z}}$ a point on the surface and expression

$$G_0 = -\frac{\exp(iK_0|\mathbf{R} - \mathbf{R}'|)}{4\pi|\mathbf{R} - \mathbf{R}'|} \tag{7}$$

for the free space Green's function. Operators M and P follow similar expressions, but with the wave number

 $K = K_0 \sqrt{\varepsilon_r}$ instead of K_0 . The scattered field can be evaluated from $\hat{\mathbf{n}} \times \mathbf{E}$ and $\hat{\mathbf{n}} \times \mathbf{H}$ at any point or in far field.

[7] Due to the short coupling phenomenon for rough surface scattering [*Maystre*, 1983], the numerical support of the unknowns $\hat{\mathbf{n}} \times \mathbf{E}$ and $\hat{\mathbf{n}} \times \mathbf{H}$ is only some wavelengths larger than the support of the right hand sides $\hat{\mathbf{n}} \times \mathbf{E}^{i}$ and $\hat{\mathbf{n}} \times \mathbf{H}^{i}$. A tapered incident beam thus bounds the problem to a finite area of the interface. However, at LGI, the size of the beam's footprint on the surface increases as γ_{0}^{-3} [Soriano et al., 2006] and the tapered beam turns useless.

[8] Therefore, we consider plane wave illumination in order to address LGI, and now following the GMoM, the surface Σ is a local perturbation of the (*xOy*) plane, with roughness over a given area. On Figure 1, *L* denotes the horizontal extension of the roughness. The idea is to find integral equations with right hand sides of numerical supports bounded to the perturbation. With an incident plane wave of wave vector $\mathbf{k}_0 - q_0 \hat{\mathbf{z}}$ and electric field

$$\mathbf{E}^{i} = \mathbf{E}_{0} e^{i(\mathbf{k}_{0} \cdot r - q_{0}z)} \tag{8}$$

of vector amplitude \mathbf{E}_0 and since the rough surface Σ is a perturbation of the (xOy) plane, it is interesting to define \mathbf{E}^r and \mathbf{E}^t the fields that would be reflected and transmitted by a plane interface $h(\mathbf{r}) = 0$. In this paper, reflected and transmitted fields will always refer to the solution of the scattering from the (xOy) plane separating vacuum and scattering medium.

$$\mathbf{E}^{r} = R \cdot \mathbf{E}_{0} e^{i(\mathbf{k}_{0} \cdot r + q_{0}z)} \quad \mathbf{E}^{t} = T \cdot \mathbf{E}_{0} e^{i(\mathbf{k}_{0} \cdot r - q_{0}'z)}$$
(9)

R and *T* are the reflection and transmission dyads built on the Fresnel coefficients. $-q'_0$ the *z* component of the transmitted wave vector $\mathbf{k}_0 - q'_0 \hat{\mathbf{z}}$. It is defined by $q'_0 = (K_0^2(\varepsilon_r - 1) + q_0^2)^{1/2}$ with $0 \le \arg q'_0 \le \pi/2$.

[9] From now on, vector field **F** stands either for the electric field **E** or the magnetic field **H**. The sole contribution from roughness to the total field in vacuum is $\mathbf{F}^s = \mathbf{F} - \mathbf{F}^i - \mathbf{F}^r$. Hereafter \mathbf{F}^s is called the scattered field, and its tangentials components $\mathbf{m} = \hat{\mathbf{n}} \times \mathbf{E}^s$ and $\mathbf{j} = \hat{\mathbf{n}} \times \mathbf{H}^s$ are solution of modified integral equations for the air

$$\left(\frac{1}{2} + M_0\right)\mathbf{j} - \frac{i}{\omega\mu_0}P_0\,\mathbf{m} = \mathbf{0} \tag{10}$$

$$\left(\frac{1}{2} + M_0\right)\mathbf{m} + \frac{i}{\omega\varepsilon_0}P_0\,\mathbf{j} = \mathbf{0} \tag{11}$$

and for the lower medium.

$$\left(\frac{1}{2} - M\right)\mathbf{j} + \frac{i}{\omega\mu_0}P\mathbf{m} = \left(\frac{1}{2} - M\right)\hat{\mathbf{n}} \times \left(\mathbf{H}^i + \mathbf{H}^r - \mathbf{H}^t\right) + \frac{i}{\omega\mu_0}P\hat{\mathbf{n}} \times \left(\mathbf{E}^i + \mathbf{E}^r - \mathbf{E}^t\right)$$
(12)

$$\begin{pmatrix} \frac{1}{2} - M \end{pmatrix} \mathbf{m} - \frac{i}{\omega \varepsilon_0 \varepsilon_r} P \mathbf{j} = \begin{pmatrix} \frac{1}{2} - M \end{pmatrix} \hat{\mathbf{n}} \times \left(\mathbf{E}^i + \mathbf{E}^r - \mathbf{E}^t \right) - \frac{i}{\omega \varepsilon_0 \varepsilon_r} P \hat{\mathbf{n}} \times \left(\mathbf{H}^i + \mathbf{H}^r - \mathbf{H}^t \right)$$
(13)

This can be proven for example by considering the integral equations satisfied by the tangential components of \mathbf{F}^i , \mathbf{F}^r

and \mathbf{F}^{t} . By definition, $\hat{\mathbf{n}} \times (\mathbf{F}^{i} + \mathbf{F}^{r} - \mathbf{F}^{t}) = \mathbf{0}$ at any point of the (xOy) plane, that is satisfying z = 0 and $\hat{\mathbf{n}} = \hat{\mathbf{z}}$. As a consequence $\hat{\mathbf{n}} \times (\mathbf{E}^{i} + \mathbf{E}^{r} - \mathbf{E}^{t})$ and $\hat{\mathbf{n}} \times (\mathbf{H}^{i} + \mathbf{H}^{r} - \mathbf{H}^{t})$ both vanish outside the rough region, and even if M and P are integral operators, we can assume that the right hand sides of (12) and (13) do show bounded support. Finally, the four integral equations (10), (11), (12) and (13) can be combined in the system

$$\begin{bmatrix} 1 + M_{0-M} & -\frac{i}{\omega\mu_0}(P_0 - P) \\ + \frac{i}{\omega\mu_0}(P_0 - P) & \frac{1 + \varepsilon_r}{2} + M_0 - \varepsilon_r M \end{bmatrix} \begin{bmatrix} \mathbf{j} \\ \mathbf{m} \end{bmatrix}$$
$$= \begin{bmatrix} \frac{1}{2} - M & +\frac{i}{\omega\mu_0}P \\ -\frac{i}{\omega\mu_0}P & \varepsilon_r(\frac{1}{2} - M) \end{bmatrix} \begin{bmatrix} \hat{\mathbf{n}} \times (\mathbf{H}^i + \mathbf{H}^r - \mathbf{H}^i) \\ \hat{\mathbf{n}} \times (\mathbf{E}^i + \mathbf{E}^r - \mathbf{E}^i) \end{bmatrix}$$
(14)

where the left-hand side matrix bears only weakly singular integral operators: M, M_0 and $P_0 - P$ [Martin and Ola, 1993]. Operators P_0 and P are hyper singular, since their kernels include the second derivative of the Green's function. They appear separately in the right-hand side of (14), but since they apply to fields with analytical expressions, one [Poggio and Miller, 1973] or two [Nédélec, 2001] derivatives can be transferred from the kernel to the fields. System (14) is designed for the numerical solution of the rough surface scattering problem at LGI, and can be discretized with low-order method of moment.

[10] Since $-q'_0$ the vertical component of the transmitted wave vector has an imaginary part for lossy dielectrics and conductors, the transmitted field is exponentially decreasing for negative heights, but also exponentially increasing for positive heights. This leads to difficulties with the numerical solving of (14) in the case of very lossy dielectrics and rough surfaces with large height root mean squares. This integral system should be restricted to transparent or slightly lossy dielectrics. For numerous media [Senior and Volakis, 1995] the total electric and magnetic fields satisfy a boundary condition on the surface that is local, or can be assumed to be so. This impedance boundary condition is of general expression

$$\hat{\mathbf{n}} \times \mathbf{E} = \hat{\mathbf{n}} \times Z_h \, \hat{\mathbf{n}} \times \mathbf{H} \tag{15}$$

with Z_h a local operator defined for the rough surface $z = h(\mathbf{r})$. It is assumed that Z_h turns to a constant Z_0 on the (xOy) plane, so that the reflected field is a plane wave for an incident plane wave and satisfy on the mean plane $(z = 0 \Rightarrow \hat{\mathbf{n}} = \hat{\mathbf{z}})$ the relation

$$\hat{\mathbf{z}} \times (\mathbf{E}^{i} + \mathbf{E}^{r}) = \hat{\mathbf{z}} \times Z_{0} \, \hat{\mathbf{z}} \times (\mathbf{H}^{i} + \mathbf{H}^{r})$$
 (16)

With (15) and (16) included in (10) or (11), two single integral equations with unknown \mathbf{j} are obtained. For example, (15) in (10) gives

$$\left\{ \left(\frac{1}{2} + M_0\right) - \frac{i}{\omega\mu_0} P_0 \hat{\mathbf{n}} \times Z_h \right\} \mathbf{j}$$

= $\frac{iP_0 \hat{\mathbf{n}} \times}{\omega\mu_0} \left\{ Z_h \hat{\mathbf{n}} \times \left(\mathbf{H}^{\mathbf{i}} + \mathbf{H}^{\mathbf{r}}\right) - \left(\mathbf{E}^{\mathbf{i}} + \mathbf{E}^{\mathbf{r}}\right) \right\}$ (17)

The relation (16) implies that the term $\hat{\mathbf{n}} \times \{Z_h \hat{\mathbf{n}} \times (\mathbf{H}^i + \mathbf{H}^r) - (\mathbf{E}^i + \mathbf{E}^r)\}$ of the right hand side vanishes outside the rough region, thus auguring a bounded right-hand side for (17) or other integral equation in the impedance case. The perfectly conducting case corresponds to the trivial $Z_h = Z_0 = 0$, while the Leontovich boundary condition writes $Z_h = Z_0 = \sqrt{\frac{\mu_0}{\varepsilon_0 \varepsilon_r}}$. Surface impedance with curvature correction for the homogeneous complex dielectric lower medium is derived from Maxwell's equations in the work of *Ong et al.* [1994] and *Marvin and Celli* [1994], and is included in a rough surface scattering model for metals in Optics and the seawater at microwave frequency in the work of *Soriano and Saillard* [2001, 2003].

[11] To conclude this section, integral equations (14) and (17) allow the numerical computation of the tangential components of the electromagnetic field in a rough surface wave scattering problem with an incident plane wave and without edge effects.

3. Scattered Field

[12] Following the boundary integral formalism in homogeneous media, the electromagnetic field at any point is related to the tangential components of the field on the interface. With $\mathbf{F}^s = \mathbf{F} - \mathbf{F}^i - \mathbf{F}^r$ for $\mathbf{F} = \mathbf{E}$, \mathbf{H} and $\mathbf{m} = \hat{\mathbf{n}} \times \mathbf{E}^s$, $\mathbf{j} = \hat{\mathbf{n}} \times \mathbf{H}^s$ and G_0 the free space Green's function for vacuum (7), it writes for the electric field

$$-\frac{\iota}{\omega\varepsilon_{0}}\operatorname{curl}\operatorname{curl}(G_{0} * \mathbf{j} \ \delta_{\Sigma}) -\operatorname{curl}(G_{0} * \mathbf{m} \ \delta_{\Sigma}) = \begin{cases} \mathbf{E}^{s}, z > h(\mathbf{r}) \\ \mathbf{0}, z < h(\mathbf{r}) \end{cases}$$
(18)

with * the space convolution operation and δ_{Σ} de Dirac function associated to the surface. As soon as fictious currents **m** and **j** are of bounded numerical support, the scattered field writes in far-field condition as a spherical wave of vector amplitude **S**. At point of the vacuum $\mathbf{R}_0 = \mathbf{r}_0 + z_0 \hat{\mathbf{z}}$, $z_0 > h(\mathbf{r}_0)$

$$\mathbf{E}^{s} = \mathbf{E} - \mathbf{E}^{i} - \mathbf{E}^{r} \sim -i \frac{e^{iK_{0}|\mathbf{R}_{0}|}}{|\mathbf{R}_{0}|} \mathbf{S}^{+}(\mathbf{k})$$
(19)

at the far-field limit for $\mathbf{K}|\mathbf{R}_0| = K_0\mathbf{R}_0$. One can define out of equation (18) the so-called vector scattering amplitude

$$\mathbf{S}^{\pm}(\mathbf{k}) = \frac{1}{4\pi} \int_{\Sigma} \left(-\mathbf{K} \times \mathbf{m}(\mathbf{R}) + \frac{1}{\omega \varepsilon_0} \mathbf{K} \times (\mathbf{K} \times \mathbf{j}(\mathbf{R})) \right)$$
$$\cdot \exp(-i\mathbf{K} \cdot \mathbf{R}) dS_{\mathbf{R}}$$
(20)

associated to the wave vector $\mathbf{K} = \mathbf{k} \pm q\hat{\mathbf{z}}$ with $\mathbf{K}^2 = \mathbf{k}^2 + q^2 = K_0^2$. The Normalized Radar Cross Section (NRCS) is proportional to the square modulus of that scattering amplitude. Since formula (18) is zero under the surface, \mathbf{S}^- vanishes for all \mathbf{k} .

$$\mathbf{S}^{-}(\mathbf{k}) = 0 \tag{21}$$

$$\mathbf{S}^{d}(\mathbf{k}) = \mathbf{S}^{+}(\mathbf{k}) - \mathbf{S}^{-}(\mathbf{k})$$
(22)

Expressions $S^{+}(k)$ and $S^{d}(k)$ are theoretically equivalent, but reveal their difference when numerically evaluated at LGS. This can be understood by comparing the term from (20) $\exp(-iqh(\mathbf{r})) \sim 1$ at grazing $q \rightarrow 0$, which turns in (22) to

$$\exp(-iqh(\mathbf{r})) - \exp(-iqh(\mathbf{r})) = -2i\sin(qh(\mathbf{r})) = O(q)$$

using Landau notation. Hence, in a numerical computation, expression $S^{+}(\mathbf{k})$ would generally lead to a nonzero value at q = 0, while expression $\mathbf{S}^{d}(\mathbf{k})$ would meet a O(q) behavior at LGS. The choice between $S^+(k)$ and $S^d(k)$ should thus be funded on theoretical arguments.

[13] Since the work by *Tatarskii and Charnotskii* [1998a, 1998b] on scalar waves, we know that the asymptotic behavior of the scattered far-field at LGI and/or LGS can be theoretically predicted. Moreover, LGI and LGS behaviors are simply connected through the Lorenz reciprocity theorem. At LGI, the limit is indicated by the right-hand side of the relevant integral equations (14) or (17). The reasoning is that integral operators and scattering formula don't depend on the incident field and q_0 . Consequently, the behavior of the right hand side is transferred to the unknowns **m** and **j** through the integral equation and then to the scattering amplitude. From the Fresnel's coefficients for a finite permittivity or conductivity, it can be easily shown that the reflection and transmission dyads $R = -I + O(q_0)$ and $T = O(q_0)$, with Landau notation, and I the dyadic identity. Then $\mathbf{E}^{i} + \mathbf{E}^{r} - \mathbf{E}^{t} = O(q_{0})$ and $\mathbf{H}^{i} + \mathbf{H}^{r} - \mathbf{H}^{t} = O(q_{0})$, with final general result that $\mathbf{S}^+(\mathbf{k}) = O(q_0 q)$, with the LGS behavior obtained by reciprocity. This holds in the impedance case, as long as $R = -I + O(q_0)$ is met. A notable exception is the perfectly conducting surface: when the permittivity is first made infinite, or equivalently $Z_h = Z_0 = 0$, the reflection coefficient for the V incident polarization is +1 for any angle. The limit $\mathbf{S}^+(\mathbf{k}) = O(q_0)$ at LGI is thus lost, and as explained by Spiga et al. [2008], the perfectly conducting model must be discarded in V polarization for incidence angles around or larger than the Brewster angle, with reciprocal situation for the V component of the scattering amplitude at LGS.

[14] Therefore, it appears that the boundary integral formalism (14) and (17) is also a theoretical tool to predict the behavior of the scattering amplitude for rough surface scattering problem at low-grazing angles. With this knowledge, one can numerically evaluate the scattering amplitude from the surface unknowns **m** and **j** through a formula that enforce its theoretical behavior at LGS. When the rough interface separates vacuum and an homogeneous medium with finite permittivity, the scattering amplitude show a $\mathbf{S}^+(\mathbf{k}) = O(q_0)$ theoretical limit at LGI. Numerically fitting this behavior with formula (20) requires accurate cancelation of contributions from all over the rough area. It will be proved in section 5 that the scattering amplitude predicted by formula (22) coincides with the more classical

(20) at non-grazing scattering angle and ensures the correct asymptotic behavior at LGS.

4. Extrapolation

[15] Solving such scattering problems, even with fast numerical solvers, remains time consuming. In addition, the iterative solvers do not take benefit from the solution at a given incidence angle to speed up the computation for another incidence angle.

[16] With the aim of saving computation time, we show how a single computation of the backscattered field for a given grazing incidence can be extrapolated from the knowledge of the theoretical asymptotic behavior of the scattering amplitude. (17) and (14) are linear integral equation or system that we symbolically denote

$$\mathbf{A}\mathbf{X} = \mathbf{B} \tag{23}$$

The integral operators, M_0 , P_0 and so on, depend on the roughness of the surface and boundary conditions, but not on the incident field. As such, **A** is not a function of γ_0 or q_0 . Equation (23) can be derived with respect to q_0 , and at the LGI limit

$$\mathbf{A}\left(\frac{\partial \mathbf{X}}{\partial q_0}\right)_{q_0=0} = \left(\frac{\partial \mathbf{B}}{\partial q_0}\right)_{q_0=0}$$
(24)

This linear equation can be solved to compute $\left(\frac{\partial \mathbf{m}}{\partial q_0}\right)_{q_0=0}$

and $\left(\frac{\partial \mathbf{j}}{\partial q_0}\right)_{q_0=0}$. [17] Since the scattering amplitude depends on q_0 only through **m** and **j**, the derivative $\left(\frac{\partial \mathbf{S}^{\pm}}{\partial q_0}\right)_{q^0=0}$ follows the same expression (20) as \mathbf{S}^{\pm} , with only **m** and **j** replaced by their LGI derivatives $\left(\frac{\partial \mathbf{m}}{\partial q_0}\right)_{q^0=0}$ and $\left(\frac{\partial \mathbf{j}}{\partial q_0}\right)_{q^0=0}$, respectively.

[18] Denoting by σ_h the RMS height of the rough surface, a first-order expansion

$$\mathbf{S} \simeq q_0 \left(\frac{\partial \mathbf{S}}{\partial q_0}\right)_{q_0=0} \tag{25}$$

is expected to be accurate as far the product $q_0\sigma_h$ remains small.

5. Numerical Results

[19] Here, the integral equation or system derived in section 2 is solved with the help of the method of moments. Basis and testing functions are pulse (piecewise constant) and Dirac functions, respectively. The linear system stemming from discretization, with its matrix full, is far too large for direct algorithms. Its solution is based on a fast iterative scheme. With the multilevel canonical grid method [Li et al., 2001; Soriano and Saillard, 2003], the matrix-vector product is performed at floating point operation (CPU) and central memory (RAM) costs that behave as $N \log_2 N$ and N, respectively, where N is the number of unknowns. The system is solved with a preconditioned GMRES.

[20] The first example aims at pointing out the advantage of the approach used here. For this purpose, we consider the sea surface described by a fully developed spectrum [Elfouhaily et al., 1997], with wind speed 3 m/s. To speed up the computation, it is assumed to be perfectly conducting.



Figure 2. HH component of the bistatic normalized scattering cross section at L band (1.2 GHz) and 89° incidence versus scattering angle in the plane of incidence for sea surface with wind speed 3 m/s. The surface is assumed to be perfectly conducting. The NRCS is derived from three different scattering amplitude expressions S^+ (solid line), S^- (dotted line) and the difference $S^d = S^+ - S^-$ (dash-dotted line).

The HH bistatic NRCS at L band (1.2 GHz, wavelength $\lambda = 25$ cm) in the plane of incidence at 89° incidence, has been computed through averaging over 100 samples, each of them being 12 m = 48λ square. The surface is sampled with a one-eighth wavelength step; an associated $4K_0$ high frequency cut-off is applied to the sea spectrum. The roughness is limited to the central part of the samples, with 2λ -long plateaus on the edges: this means with notations of Figure 1 that $L_g = L_d = 2\lambda$ and $L = 44\lambda$. Note also that a smooth transition is applied between the roughness and the plateaus, over 4 λ . As a consequence, the area A for the normalization of the RCS is difficult to define for the GMoM. Here, it is set to $A = 40\lambda \times 40\lambda$. This may have a little impact on comparisons to other approaches or experimental data. All those dimensions are chosen with no reference to the incidence angle. Still, for the Monte Carlo average to be statistically significant, the length of the rough part of the samples has to be much larger than the characteristic horizontal scales of the roughness, that is the wavelength of the peak wave in the ocean surface case. Three curves are plotted in Figure 2: one represents the bistatic NRCS as derived from Stratton-Chu formula (S^+) , one is the NRCS in the lower medium, which, according to the extinction theorem, should vanish (S⁻) and the third one is the NRCS derived from the difference $S^d = S^+ - S^-$. As usual, the angular resolution of those diagrams is inversely proportional to the samples dimensions. In order for rapid angular variations in radar cross sections to be studied, one has to

correctly set the surface size, and to test convergence on this parameter.

[21] Clearly, the use of Stratton-Chu formula S^+ does not provide an accurate estimation of the bistatic cross section beyond -75° . The ratio between the largest and the smallest value of the NRCS, which can be seen as the numerical dynamic range, does not exceed 10^6 . It is linked to the accuracy with which the extinction theorem S^- is satisfied. Performing the difference (22) permits to significantly improve the dynamic range and to compute the NRCS at grazing scattering angles.

[22] We now test the performances of first orders approximate surface scattering methods at low-grazing angles. Since the Kirchhoff approximation of Beckmann and Spizzichino [1963] fails to show the correct qq_0 behavior at grazing, it is discarded and we focus on the small perturbation method (SPM1) of Rice [1951] and the small slope approximation (SSA1) of Voronovich [1994]. There exist for such theories statistical formulations associated to a plane wave illuminating an infinite surface with stationary roughness. In order to avoid Monte Carlo and normalization errors, and to focus on the impact of physical assumptions on the scattering at grazing, approximate models NRCS are computed through Monte Carlo average, with the same samples and configuration as for the GMoM. Still considering the ocean surface at L band and 89° incidence, we compare on Figure 3 the bistatic diagrams of the GMoM and SSA1. The standard validity criterions are $\sigma_h < \lambda/20$ for SPM1 and $K_0 \sigma_h \sigma_s < 1$ for SSA1, with σ_h and σ_s the surface



Figure 3. HH component of the bistatic normalized scattering cross section predicted by the GMoM (dashed line) and SPM1 or SSA1 (solid line) at L band (1.2 GHz) and 89° incidence versus scattering angle in the plane of incidence for sea surface with wind speed 3 m/s. The surface is assumed to be perfectly conducting.

height and slope root mean squares. Our surface is outside SPM1's validity domain, but inside SSA1's, with $K_0\sigma_h\sigma_s = 0.16$. Nevertheless, on this example, the SPM1 diagram do coincide perfectly with SSA1 on the whole scattering angles range, and thus is not plotted. It can be seen that the approximate models underestimate the NRCS by several dBs, with the error increasing from 2 dB at +80° to 8 dB in low-grazing backward angles and in the backscattering direction.

[23] To extend this study, we now consider rough surfaces with a Gaussian spectrum of correlation radius $\lambda/2$ and a height root mean square varying from 0.08λ to 0.83λ , in the perfectly conducting case. With such a short correlation radius, the sample side can be shortened to 16λ , with 2λ plateaus, still. The number of samples for the Monte Carlo average is now 200. In Figure 4, the HH monostatic NRCS difference between SSA1 and the GMoM is plotted against the surfaces height root mean square for angles 20°, 60°, 80° and 89°. If the SPM1 enjoys a statistical expression for its NRCS, the GMoM mean NRCS is obtained through Monte Carlo average on a finite number of samples. Corresponding error bars appear on plots of Figure 4. At the small height limit, the curves for all angles tend toward zero, thus underlining the accuracy of the GMoM. Then, the SSA1 error increases more or less linearly with the height root mean square, but with slopes that strongly depend on the angle. For a height of $\lambda/20$, its error is smaller that 1 dB 20° angle, around 3 dB at 60°, but reaches 8 dB at 80° and even 10 dB at 89°. For that Gaussian surfaces, SPM1 and SSA1 do not coincide exactly, still the results and conclusions are similar. The GMoM clearly states that the domain of validity of approximate methods varies with the angle and narrows at grazing. Beyond 80°, the 3 dB error line is crossed as soon as $\lambda/60$ height root mean square for both SPM1 and SSA1.

[24] Up to now, bistatic diagrams have been preferently presented, since with the Method of Moment, for a given incidence angle, it is easy and fast to compute through formula (20) the scattered field for any (reasonable) number of scattering angles, but when the incidence is varied, the integral equation (17) or system (14) is to be solved for a new right-hand side. When direct methods such as LU decomposition are used to solve the linear system that comes from the discretization of integral equations, multiple right-hand sides are not a problem. However, the number of unknowns N is too large for today's computers, except for one-dimensional (1D) surfaces, i.e. with an invariance along one horizontal direction. Beyond this assumption, the system is solved iteratively, and the computing time is



Figure 4. HH monostatic NRCS error in the plane of incidence between SPM1 and the GMoM versus the surfaces height root mean square for perfectly conducting surfaces with Gaussian spectrum of correlation radius $\lambda/2$ and angles 20°, 60°, 80° and 89°, with error bars for the GMoM Monte Carlo average.



Figure 5. HH bistatic NRCS for incidence angles 60° , 80° , 87° and 89° (solid lines) and extrapolated HH monostatic NRCS (dashed line) for perfectly conducting surfaces with Gaussian spectrum of height root mean square 0.83λ and correlation radius $\lambda/2$ versus scattering angle in the plane of incidence.

proportional to the number of incidence angles. For the monostatic diagram in the low-grazing angles region, one can try extrapolation (25). On Figure 5 is plotted for the rougher Gaussian surface the bistatic NRCS at backward scattering angles for incidence angles 60° , 80° , 87° and 89° , and the NRCS obtained through (25). It appears that extrapolation matches the rigorous computation at angles 89° and 87° . At 80° , a discrepancy starts being visible, but the error is smaller than 1 dB, and obviously the extrapolation looses accuracy at larger grazing angles, with a 3.5 dB error at 60° .

6. Conclusion

[25] The incidence angle is no longer a limit to the numerical solution of the rough surface electromagnetic wave scattering problem. We have presented a grazing surface integral formalism with a numerical cost made independent of the incidence. The cases of a rough interface between two homogeneous media as well as the impedance boundary condition are detailed in full vector theory. In this paper, the model has been implemented following the Method of Moments with pulse basis functions and Dirac testing functions, a classical scheme. As a matter of fact, the integral equations are modified, but the integral operators being left untouched, the countless associated numerical techniques published in the literature for discretization, such as the Boundary Element Method [*Rao et al.*, 1982; *Chew et al.*, 2008], or acceleration techniques such as the Forward

Backward Spectral Acceleration of *Torrungrueng and Johnson* [2001] or the Fast Multipole Method of *Jandhyala et al.* [1998], might be used as is.

[26] Also, the scattering amplitude formula shows an alternative equivalent expression that reveals best fitted to low-grazing scattering angles. The presented surface integral formalism combined with the equivalent scattering formula constitute a dedicated model for the wave scattering from rough surfaces at low-grazing incidence and scattering angles.

[27] Tested first-order approximate methods, namely the small perturbation method and the small slope approximation perform very poorly at low-grazing angles, with their validity domain drastically narrowed. At grazing, multiple scattering is often modelized by a shadowing function [*Bourlier et al.*, 2002]. However, it appears that first-order approximations systematically underestimate the Radar Cross Section at grazing angles, and thus applying a shadowing function would only increase the gap between rigorous and approximate computations.

[28] Nevertheless, the grazing angles regime also have positive characteristics. It has been shown how the monostatic diagram at large incidence angles can be extrapolated from a single (statistical) calculus at grazing incidence. This is a lead for the development of approximate scattering models devoted to the low-grazing angles.

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M. Saillard, Laboratoire de Sondage Electromagnétique de l'Environnement Terrestre, Université du Sud Toulon-Var, F-83957 La Garde, CEDEX, France. (saillard@univ-tln.fr)

G. Soriano, Sondage ElectroMagnétique et Optique, Institut Fresnel, Aix-Marseille Université, Campus de St Jérôme, F-13397 Marseille, CEDEX, France. (gabriel.soriano@fresnel.fr)